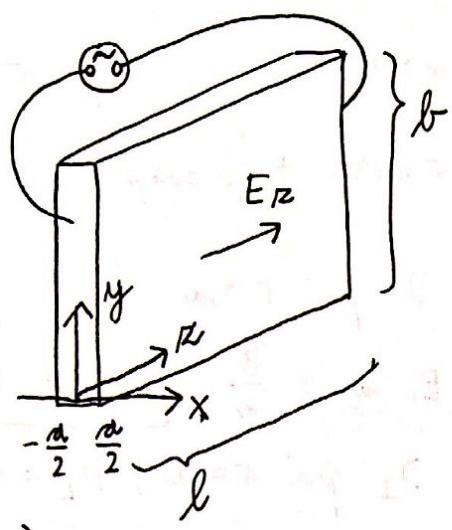


2. KOLOKVIJ

1 KOŽNI POJAV V TRAKU



$$\begin{aligned} \vec{\nabla} \times \vec{E} &= -\frac{\partial \vec{B}}{\partial t} \\ \vec{\nabla} \times \vec{B} &= \mu_0 \vec{j} = \mu_0 \sigma \vec{E} \end{aligned} \Rightarrow \vec{\nabla} \times (\vec{\nabla} \times \vec{E}) = -\mu_0 \sigma \frac{\partial \vec{E}}{\partial t}$$

$$\nabla^2 \vec{E} - \mu_0 \sigma \frac{\partial \vec{E}}{\partial t} = 0$$

$$\vec{E} = \vec{E}(\vec{r}) e^{i\omega t} \Rightarrow \nabla^2 \vec{E} - i\mu_0 \sigma \omega \vec{E} = 0$$

$$k = (1+i) \sqrt{\frac{\mu_0 \sigma \omega}{2}} \leftarrow k^2$$

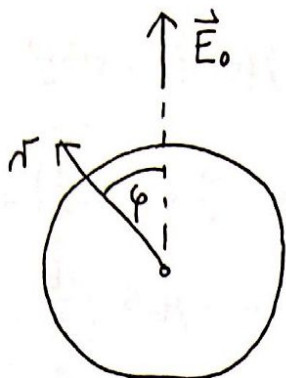
1/4 a) - ravina nos E_z , $E_z = E_z(x)$, saj sta dimenziji y & z veči
 $\hookrightarrow \nabla^2 = \frac{\partial^2}{\partial x^2}$
 $E_z'' - k^2 E_z = 0 \rightarrow E_z(x) = A \operatorname{ch} kx$, saj mora biti SODA funkcija zaradi simetrije

+ - robni pogoj: $E_z(\pm \frac{a}{2}) = A \operatorname{ch} \frac{ka}{2} = E_0 \Rightarrow A = \frac{E_0}{\operatorname{ch} \frac{ka}{2}}$
 $E_z(x) = E_0 \frac{\operatorname{ch} kx}{\operatorname{ch} \frac{ka}{2}}$

- celoten električni tok:
 1/4 $I = \sigma \int E_z(x) dS = \sigma b \int_{-a/2}^{a/2} E_z(x) dx = \frac{\sigma b E_0}{\operatorname{ch} \frac{ka}{2}} \int_{-a/2}^{a/2} \operatorname{ch} kx dx =$
 $= \frac{\sigma b E_0}{k \operatorname{ch} \frac{ka}{2}} 2 \operatorname{sh} \frac{ka}{2} = \sigma a b E_0 \frac{2}{ka} \operatorname{th} \frac{ka}{2}$
 + $Z = \frac{U}{I} = \frac{E_0 l}{I} = \frac{l}{\underbrace{\sigma a b}_{R_0} \operatorname{th} \frac{ka}{2}} = \boxed{R_0 \frac{\frac{ka}{2}}{\operatorname{th} \frac{ka}{2}}}$

b) visoke $\omega \rightarrow \frac{ka}{2} \gg 1 \Rightarrow \operatorname{th} \frac{ka}{2} = 1 \Rightarrow Z = R_0 \frac{ka}{2}$
 + $R = \operatorname{Re}(Z) = R_0 \frac{a}{2} \sqrt{\frac{\mu_0 \sigma \omega}{2}} \Rightarrow \boxed{\frac{R}{R_0} = \sqrt{\frac{\mu_0 \sigma \omega}{2}} \frac{a}{2}}$ 1

2) DIELEKTRIČNI VALJ V HOMOGENEM POLJU



$$\nabla^2 U(r, \varphi) = 0$$

$$\frac{1}{4} \left\{ U(r, \varphi) = \begin{cases} A r \cos \varphi & , r < a \\ -E_0 r \cos \varphi + \frac{B}{r} \cos \varphi & , r > a \end{cases} \right.$$

+ { RP1: zvezanost U pri $r=a$

$$A a = -E_0 a + \frac{B}{a}, \quad A = -E_0 + \frac{B}{a^2}$$

+ { RP2: zvezanost D_{\perp} pri $r=a$, $D_{\perp} = -\epsilon \frac{\partial U}{\partial r}$

$$-\epsilon A = E_0 + \frac{B}{a^2}$$

$$\frac{1}{4} \left\{ (RP1) - (RP2) : A(1+\epsilon) = -2E_0 \right.$$

$$A = -\frac{2}{\epsilon+1} E_0$$

- konstruiraj valja: $U(r, \varphi) = -\frac{2}{\epsilon+1} E_0 \overbrace{r \cos \varphi}^z$

$$\frac{1}{4} \left\{ E_1 = -\frac{\partial U}{\partial r} = \frac{2}{\epsilon+1} E_0 \rightarrow \text{homogeno polje}$$

$$P = \epsilon_0 (\epsilon - 1) E_1 = \boxed{\epsilon_0 \frac{2(\epsilon - 1)}{\epsilon + 1} E_0} \rightarrow \text{homogena polarizacija}$$

1

- alternativna pot:

$$(RP1) + (RP2) : (1-\epsilon)A = \frac{2B}{a^2} \rightarrow B = -\frac{\epsilon-1}{2} a^2 \left(-\frac{2}{\epsilon+1}\right) E_0$$

$$B = \frac{\epsilon-1}{\epsilon+1} a^2 E_0$$

$$\sigma_r = \epsilon_0 (E_{\perp}^{\text{zUN}} - E_{\perp}^{\text{NOT}}) = \epsilon_0 \left(E_0 + \frac{B}{a^2} + A \right) \cos \varphi =$$

$$= \epsilon_0 \frac{\epsilon-1-2+\epsilon+1}{\epsilon+1} E_0 \cos \varphi = \epsilon_0 \frac{2(\epsilon-1)}{\epsilon+1} E_0 \cos \varphi \Rightarrow P = \boxed{\epsilon_0 \frac{2(\epsilon-1)}{\epsilon+1} E_0}$$

$$\sigma_r = \vec{P} \cdot \vec{n} = P \cos \varphi$$

3 KRIŽNA ANTENA

$\left. \begin{array}{l} \text{a) } \text{ena prečka: } p_e = el \\ \dot{p}_e = \dot{i}l = Il \\ \ddot{p}_e = \ddot{i}l \end{array} \right\} \text{cela antena: } \begin{array}{l} I_1 = I_0 \cos \omega t \\ I_2 = I_0 \sin \omega t \\ \downarrow \\ \ddot{p}_{e1} = \ddot{i}_1 l \hat{e}_x = -I_0 l \omega^2 \sin \omega t \hat{e}_x \\ \ddot{p}_{e2} = \ddot{i}_2 l \hat{e}_y = I_0 l \omega^2 \cos \omega t \hat{e}_y \end{array}$

$$\ddot{p}_e = I_0 l \omega^2 [\hat{e}_y \cos \omega t - \hat{e}_x \sin \omega t] \Leftarrow$$

$\vec{B} = -\frac{\mu_0}{4\pi r} \hat{e}_r \times \ddot{p}_e(t_r) =$

$$= -\frac{\mu_0 I_0 l \omega^2}{4\pi r} [\hat{e}_r \times \hat{e}_y \cos \omega t_r - \hat{e}_r \times \hat{e}_x \sin \omega t_r]$$

$\hat{e}_r = \begin{bmatrix} \cos \varphi \sin \vartheta \\ \sin \varphi \sin \vartheta \\ \cos \vartheta \end{bmatrix}, \quad \hat{e}_r \times \hat{e}_y = \begin{bmatrix} -\cos \vartheta \\ \vartheta \\ \cos \varphi \sin \vartheta \end{bmatrix}, \quad \hat{e}_r \times \hat{e}_x = \begin{bmatrix} \vartheta \\ \cos \vartheta \\ -\sin \varphi \sin \vartheta \end{bmatrix}$

$\vec{E} = -\hat{e}_r \times c_0 \vec{B}$

$$\vec{P} = \frac{1}{\mu_0} \vec{E} \times \vec{B} = -\frac{c_0}{\mu_0} (\hat{e}_r \times \vec{B}) \times \vec{B} = \frac{c_0}{\mu_0} B^2 \hat{e}_r$$

$$\vec{B} (\hat{e}_r \cdot \vec{B}) - \hat{e}_r B^2$$

$B^2 = \frac{\mu_0^2 I_0^2 l^2 \omega^2}{16\pi^2 c_0^2 r^2} \left[(\cos^2 \vartheta + \cos^2 \varphi \sin^2 \vartheta) \overset{\langle \rangle}{\cos^2 \omega t_r} + (\cos^2 \vartheta + \sin^2 \varphi \sin^2 \vartheta) \overset{\langle \rangle}{\sin^2 \omega t_r} + \overset{\vartheta}{\uparrow \langle \rangle} \cos \omega t_r \sin \omega t_r \right]$

$\frac{1}{4} \langle P \rangle = \frac{\mu_0 I_0^2 l^2 4\pi^2 c_0^2}{16\pi^2 c_0 r^2 \lambda^2} \left[(\cos^2 \vartheta + \cos^2 \varphi \sin^2 \vartheta) \frac{1}{2} + (\cos^2 \vartheta + \sin^2 \varphi \sin^2 \vartheta) \frac{1}{2} + \vartheta \right] =$

$$= \frac{Z_0 I_0^2 l^2}{4\pi^2 \lambda^2} \cdot \frac{1}{2} (2 \cos^2 \vartheta + \sin^2 \vartheta) = \frac{Z_0 I_0^2}{4\pi^2} \left(\frac{l}{\lambda}\right)^2 \cdot \frac{1}{2} (1 + \cos^2 \vartheta)$$

$$+ \left\{ \langle \int P dS \rangle = \frac{Z_0 I_0^2}{4} \left(\frac{l}{\lambda}\right)^2 2\pi \int d(\cos\vartheta) \frac{1}{2} (1 + \cos^2\vartheta) = \frac{2\pi}{3} Z_0^2 I_0^2 \left(\frac{l}{\lambda}\right)^2 \right.$$

$$\left. \frac{1}{2} \cdot \left(2 + \frac{2}{3}\right) = \frac{4}{3} \right.$$

8) primer Hertzovega dipola → le eden izmed členov v izrazu za \vec{B} → recimo \hat{e}_y

$$+ \left\{ B^2 = \frac{\mu_0^2 I_0^2 l^2 \omega^2}{16\pi^2 \epsilon_0^2 r^2} (\cos^2\vartheta + \cos^2\varphi \sin^2\vartheta) \cos^2\omega t r \right.$$

$$\downarrow$$

$$\langle P \rangle = \frac{Z_0 I_0^2 l^2}{4r^2 \lambda^2} \cdot \frac{1}{2} (\cos^2\vartheta + \cos^2\varphi \sin^2\vartheta)$$

$$+ \left\{ \langle \int P dS \rangle = \frac{Z_0 I_0^2}{4} \left(\frac{l}{\lambda}\right)^2 \frac{1}{2} \int d\varphi d(\cos\vartheta) (\cos^2\vartheta + \cos^2\varphi \sin^2\vartheta) \right.$$

$$\left. 2\pi \cdot \frac{2}{3} + \frac{1}{2} 2\pi \left(2 - \frac{2}{3}\right) = \frac{4\pi}{3} + \frac{4\pi}{3} = \frac{8\pi}{3} \right.$$

$$\boxed{\langle \int P dS \rangle = \frac{\pi}{3} Z_0^2 I_0^2 \left(\frac{l}{\lambda}\right)^2} \rightarrow \text{DVAKRAT manj, kot za krivins anteno!} \quad \boxed{1}$$

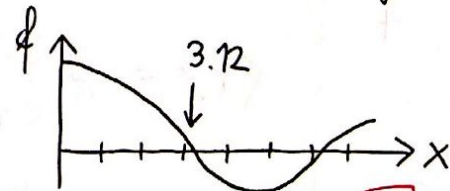
Rezultat je rdi intuitiven, a ni trivialen. Primer: za dve enaki vzporedni anteni je moč 4-krat večja, če sta v fazi, in enaka 0, če sta v nasprotni fazi.

4 KOAKSIALNI KABEL → TM način

$$E_r(r, \varphi) = R(r) \Phi(\varphi), \quad \text{Besslova enačba za } R(r)$$

$$+ \left\{ R(r) = A_m J_m(\alpha r) + B_m N_m(\alpha r) \right. \rightarrow \text{Neumannove funkcije so v koaksialnem kabelu možne!}$$

$$\text{RP: } \left\{ \begin{aligned} R(a) &= A_m J_m(\alpha a) + B_m N_m(\alpha a) = 0 \\ R(2a) &= A_m J_m(2\alpha a) + B_m N_m(2\alpha a) = 0 \end{aligned} \right.$$



$$\frac{1}{4} \left\{ \begin{aligned} \Phi(x) & \times \begin{bmatrix} J_m(\alpha a) & N_m(\alpha a) \\ J_m(2\alpha a) & N_m(2\alpha a) \end{bmatrix} \begin{bmatrix} A_m \\ B_m \end{bmatrix} = 0 \Rightarrow \det = 0 \\ J_m(\alpha a) N_m(2\alpha a) - J_m(2\alpha a) N_m(\alpha a) &= 0 \end{aligned} \right. \rightarrow \left. \begin{aligned} m=0, \text{ min ničla} \\ \alpha a = \frac{\omega_{\min}}{c_0} a = 3.12 \end{aligned} \right\} \frac{1}{2}$$